A Blind Antenna Selection Scheme for Single-Cell



Uplink Massive MIMO



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Abstract	2.1.2 Greedy Approach-based Antenna Selection	3.2 MSE Performance
We approach the problem of antenna selection in single-cell uplink massive MIMO using two different techniques. The first one consists in solving a convex relaxation of the problem using standard convex optimization tools. The second technique solves the problem using a greedy approach. The main advantages of the greedy approach lies in its wider scope, in that, unlike the first approach, it can be applied irrespective of the considered performance criterion. In the case where we don't have perfect channel knowledge, We extend both approaches	0: Initialize $S = randsample(n, k)$ 0: Compute MSE* = $f(\mathcal{H}, S)$ 1: for $i = 1$ to $\#$ iterations do 1: $\overline{S} = \{1, \dots, n\} \setminus S$ 1: $j \leftarrow 1$ 2: while $j \leq n - k$ do 2: $p \leftarrow \overline{S}[j]$ 2: $\mathcal{I} \leftarrow S$ 2: table $\leftarrow \operatorname{zeros}(k, 1)$	d = 1 $d = 1$ $d = 2$ $d =$

statistics.

1. System Description

to perform blind antenna selection that is only based on the channel

1.1 Downlink Model

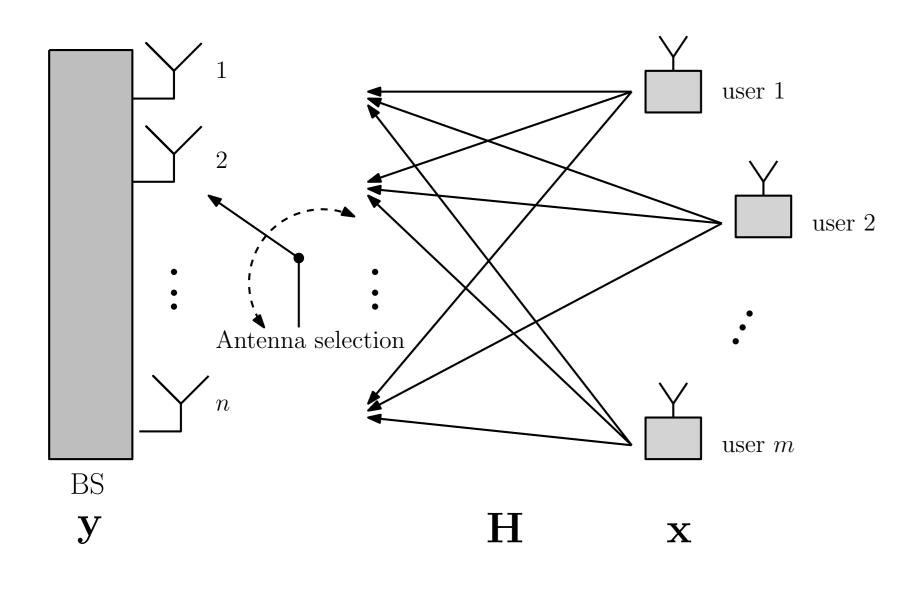


Figure 1: System model of an uplink MU-MIMO system composed of a BS equipped with n antennas and serving *m* single-antenna users.

The received signal vector at the BS is given by

- 3: for $l = 1 \rightarrow k$ do $\mathcal{I}\left[l\right] \leftarrow p$ 3: $\mathsf{table}\left[l\right] \leftarrow f\left(\mathcal{H}, \mathcal{I}\right)$ 3: $\mathcal{I} \leftarrow \mathcal{S}$ 3:
- end for 4:
- if $\min(\text{table}) < MSE^*$ then 5:
- $MSE^* \leftarrow \min(table)$ 5:
- $\mathcal{S}[\arg\min(\mathsf{table})] \leftarrow p$ 5:
- end if 6:
- end while 7:
- 8: end for

Algorithm 1: Greedy Approach for Antenna Selection

2.2 Blind Antenna Selection

Assumption 1 We assume that *n*, *m* and *k* grow simultaneously large while 1. $\frac{n}{m} \to c \in (1, \infty)$

2. $0 < \liminf \frac{k}{n} < \limsup \frac{k}{n} < 1.$

3. $\liminf \frac{k}{m} > 1.$

Assumption 2 The correlation matrix R satisfies

1. $\sup_n \|\mathbf{R}\| < \infty$

 $2.\inf_n \frac{1}{n}\operatorname{tr} \mathbf{R} > 0.$

(1)

(2)

(3)

(4)

(6)

(7)

Lemma 1 [1] Let Assumptions 1 and 2 hold true. Let δ be the unique solution to the following equation

$$\delta = m \left(\operatorname{tr} \left[\operatorname{\mathbf{R}diag}(s) \left(\mathbf{I}_n + \delta \operatorname{\mathbf{R}diag}(\mathbf{s}) \right)^{-1} \right] \right)^{-1}$$
(8)

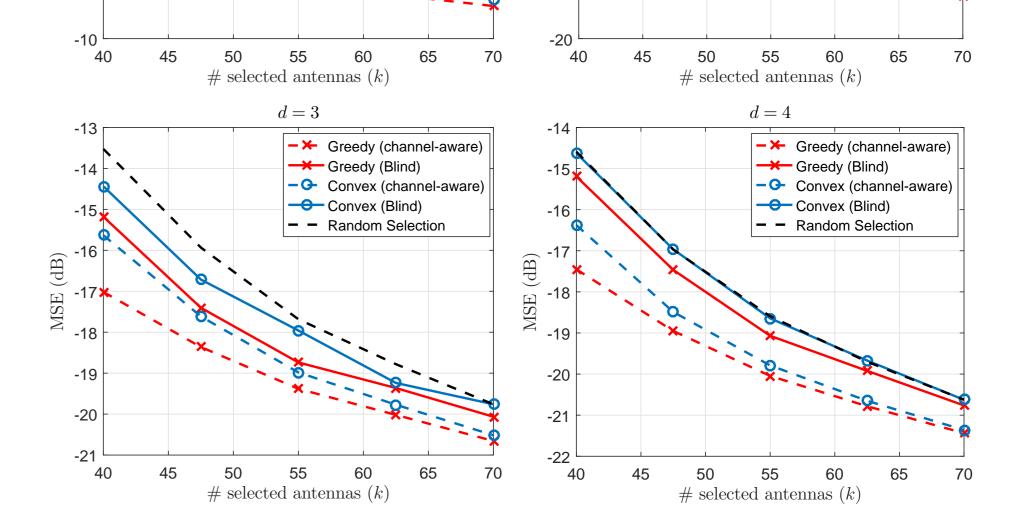
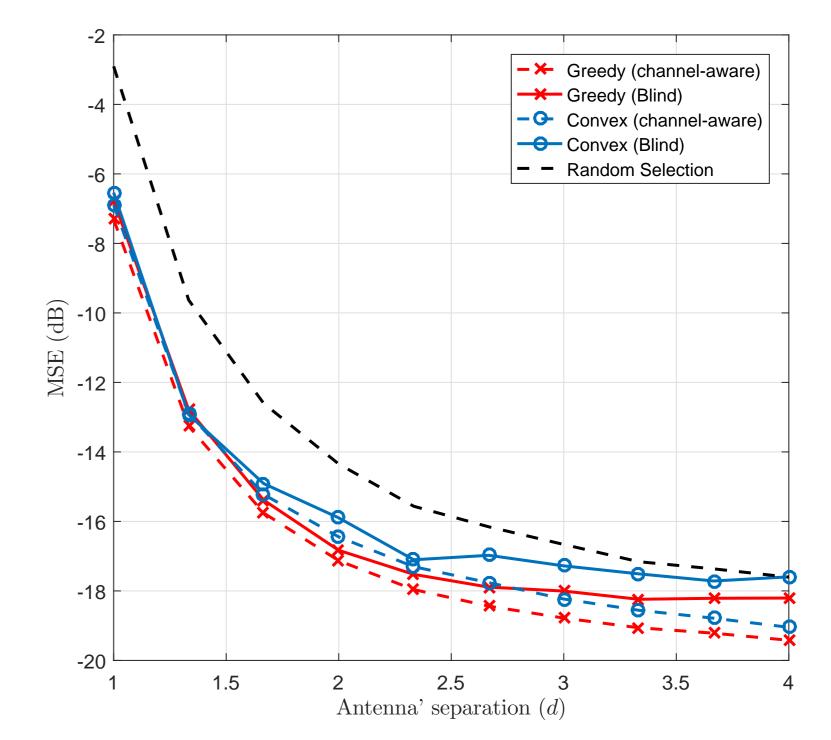


Figure 2: Average MSE achieved by the proposed selection techniques versus k for different values of the antennas' separation d.



$$\mathbf{y} = \sqrt{\rho} \mathbf{H} \mathbf{x} + \mathbf{e}.$$

The random channel H exhibits the one-sided Kronecker model given by

 $\mathbf{H} = \mathbf{R}^{\frac{1}{2}}\mathbf{W},$

where W is a matrix with *i.i.d*, CN(0, 1) entries. At the output of the zero-forcing receiver (ZF), the estimated signal is given by

$$\begin{split} \widehat{\mathbf{x}} &= \frac{1}{\sqrt{\rho}} \left(\mathbf{H}^{H} \mathbf{H} \right)^{-1} \mathbf{H}^{H} \mathbf{y} \\ \mathsf{MSE} &= \mathbb{E} \left[\| \widehat{\mathbf{x}} - \mathbf{x} \|^{2} \right] \\ &= \frac{1}{\rho} \operatorname{tr} \left[\left(\mathbf{W}^{H} \mathbf{R} \mathbf{W} \right)^{-1} \right]. \end{split}$$

2. Antenna Selection

2.1 CSI-aware Antenna Selection

Let

 $\mathbf{MSE}(\mathbf{s}) = \operatorname{tr} \left| \left(\mathbf{H}^{H} \operatorname{diag}(s) \mathbf{H} \right)^{-1} \right|$

Define $\overline{\mathbf{MSE}}(\mathbf{s})$ as $\overline{\mathbf{MSE}}(\mathbf{s}) = \delta$

Then, MSE(s) satisfies

 $\mathbf{MSE}(\mathbf{s}) - \overline{\mathbf{MSE}}(\mathbf{s}) \xrightarrow[n \to \infty]{a.s.} 0.$

MSE(s) only depends on the channel statistics **R**. **Theorem 1** 1. The function

$$\overline{\mathbf{MSE}} : \mathbb{R}^n_+ \to \mathbb{R}_+$$
$$\mathbf{s} \mapsto \overline{\mathbf{MSE}}(\mathbf{s})$$

is convex in
$$\mathbb{R}^{n}_{+}$$
.
2.

$$\frac{\partial \overline{\mathbf{MSE}}(\mathbf{s})}{\partial s_{i}} = \frac{\delta \left[\mathbf{R}^{\frac{1}{2}} (\mathbf{I} + \delta \mathbf{R} \operatorname{diag}(\mathbf{s}))^{-2} \mathbf{R}^{\frac{1}{2}} \right]_{i,i}}{\operatorname{tr} \left[\mathbf{R} \operatorname{diag}(\mathbf{s}) (\mathbf{I} + \delta \mathbf{R} \operatorname{diag}(\mathbf{s}))^{-2} \right]}.$$
(9)

Corollary 1 Convex optimization techniques can be applied to solve problems with an objective given by $\overline{MSE}(s)$. As a matter of fact, the blind antenna selection problem can be formulated as follows

$$\widehat{\mathbf{s}}_{blind} = \underset{\mathbf{s}}{\operatorname{arg\,min}} \quad \overline{\mathbf{MSE}}(\mathbf{s})$$
s.t.
$$\mathbf{1}^{T}\mathbf{s} = k$$

$$s_{i} \in \{0, 1\}, \ i = 1, \cdots, n.$$
(10)

Similarly, (10) can be solved using both heuristics: the convex relaxation and the greedy approach.

Figure 3: Average MSE achieved by the proposed selection techniques versus the antennas' separation d with k = 50.

Comments

- When the correlation between antennas is low (d = 4), the proposed blind algorithms are not that advantageous as compared to the random selection algorithm.
- However, with the impact of correlation becoming more important $(d \downarrow)$, the gain of blind approaches over the random selection approach increases.
- Blind algorithms perform antenna selection at the pace of the variation of the large scale statistics. A high reduction in the computational complexity is thus achieved compared to channel-aware algorithms.
- From a practical point of view, blind selection algorithms are more suitable since they consider practical issues such as antenna synchronization and adaptation.

4. Conclusion



Then, the selection problem is formulated as follows

minimize MSE(s) $\mathbf{1}^T \mathbf{s} = k$ s.t. $s_i \in \{0, 1\}, i = 1, \cdots, n.$

Antenna Selection via Convex Optimization 2.1.1

It is mainly based in relaxing the boolean constraints in (6)

$$\mathbf{s}_{0} = \underset{\mathbf{s}}{\operatorname{argmin}} \quad \mathbf{MSE}(\mathbf{s})$$

s.t.
$$\mathbf{1}^{T}\mathbf{s} = k$$
$$0 \le s_{i} \le 1, i = 1, \cdots, n.$$

3. Numerical Results and Discussion

m = 30 users, n = 100 antennas and $\rho = 20$ dB. We consider the correlation model given by

 $\mathbf{R}_{i,j} = \exp\left[-0.05.d^2 (i-j)^2\right], \ 1 \le i, j \le n.$ (11)

3.1 Complexity

Algorithm	Complexity
Convex Optimization(Channel-aware)	$N \times \mathcal{O}(n^3)$
Convex Optimization(Blind)	$\mathcal{O}\left(n^{3}\right)$
Greedy(Channel-aware)	$K \times N \times \mathcal{O}(n^2)$
Greedy(Blind)	$K \times \mathcal{O}\left(n^2\right)$

Table 1: Computational complexity of the different proposed algorithms.

In this work, we showed that using tools from random matrix theory, it is possible to asymptotically approximate the MSE. As such, perfect knowledge of the channel matrix is not needed and only statistics are required to perform selection. We proposed two techniques: the first is based on a greedy approach and the second is based on a convex relaxation heuristic. Numerical results showed that the blind techniques have a comparable performance to techniques that require full knowledge of the channel matrix, especially at high correlation.

5. References

[1] J. W. Silverstein and Z. D. Bai, "On the Empirical Distribution of Eigenvalues of a Class of Large Dimensional Random Matrices," Journal of Multivariate Analysis, vol. 54, pp. 175-192, May 2002.